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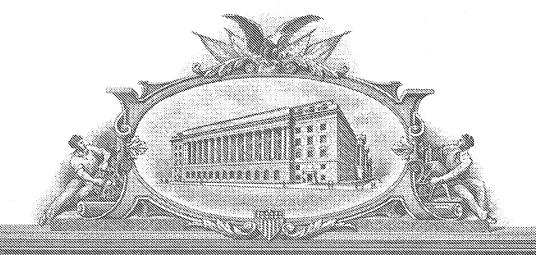
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APPLICATION NUMBER: 60/550,463

FILING DATE: March 04, 2004

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### PROVISIONAL APPLICATION FOR PATENT COVER SHEET

This is a request for filing a PROVISIONAL APPLICATION FOR PATENT under 37 CFR 1.53(c)

Assistant Commissioner for Patents Alexandria, VA 22313 BOX PROVISIONAL APPLICATION

TELEPHONE NO.: (858) 845-8355

Attorney Docket No.: 040257P1

Date: March 4, 2004

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Zhi-Zhong		Yu			San Diego, California		
Additional inventors a	re being name	ed on the sheet	attached he	reto			
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Country:	USA			: (858) 651-440			
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☐ A check or money order is enclosed to cover the filing fees ☐ The Commissioner is hereby authorized to charge filing							
The invention was made by an agency of the United States Government or under a contract with an agency of the United States							
Government.							
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SIGNATURE: DATE: March 4, 2004							

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TYPED or PRINTED NAME/REGISTRATION NO.: Timothy F. Loomis, Registration No.: 37,383



### More Detailed Description of Invention Disclosure: -

### Method to Provide a Smooth transition between adjacent GMSK and 8PSK bursts (EGPRS/EDGE)

### Describe the problem solved and the advantage(s) of the invention:

Defined by ETSI GSM 05.04, EDGE (EGPRS) uses two modulation schemes, GMSK and 8PSK. The two modulation schemes are different in many ways. The most obvious one is that GMSK has constant amplitude, while 8PSK has variable amplitude. GMSK only modulates the phase and keeps the amplitude constant, while 8PSK modulates both phase and amplitude. By doing this 8PSK triples GMSK transmitting data rate.

The introduction of EDGE (and GPRS) means in the near future will be able to make a voice call while the MS is doing data call. That often requires both GMSK and 8PSK modulations in time slot next to each other. This means that both modulation schemes can be in the same spectrum and can appear in two adjacent bursts on both downlink and uplink. For downlink Tx BS normally does not switch off (ramp down) at end of each burst, as it needs to Tx in the next burst. For uplink, when there are two adjacent time slots, similar to BS, it is desirable not to power down at the end of the first burst and power up at the beginning of the next one. As the power is on in the guard period (GP), it has to be controlled carefully to minimise the interference to others. Spec has defined a spectrum mask for the transition. It needs joint efforts from both baseband and RF to satisfy the requirement and it is desirable to keep the emission as low as possible. This proposal concerns the baseband technique which provide necessary condition to achieve the goal.

If both of the bursts are of the same modulation, the transition can be smoothly made as what BS always does for GSM, even with different power level. If the two bursts are of different modulations, i.e. GMSK followed by 8PSK, or the other way, the issue arising from such case is that the direct transition between the two modulations often generates spurious spectrum from the output stage of baseband signal, which will appear in RF and cause violation of the mask. Therefore it is necessary condition for both BS and MS baseband to be able to handle transition between adjacent GMSK and 8PSK bursts without generating unwanted frequency components. Currently there is no such application a/v in the real network around world, but in the near future, when EDGE is rolled out, user may want to make a voice call while surf the web (DTM). It is a practical issue and technical challenge for the new service in the near future. ETSI did not mention anything about the G-8 transition method. It only provide the masks to be satisfied.

Direct Interpolation between two modulation bursts might be used to make the transition smooth. However spurious spectrum could not be effectively controlled. Basically interpolation introduces a third modulation, which is neither GMSK nor 8PSK. It would be ideal not to introduce any other modulations during GP.

We are looking into this issue and find the way to make the smooth transition without using any other modulations. There are specific codes for each modulation and specific time of switching to make the transition smooth. The advantage is that there is no foreign modulation involved and there is no added complexity to the HW. These codes and the time to start their modulation coupled with the switching time are found from over 30 million combinations of code through both modulations.

### Describe how others have solved the stated problem:

There is no such service that encounter this issue at moment, however ETSI does imply such application could happen and indeed is a good feature and selling point in the foreseeable future

Traditional way of dealing with such issue would be either of following approaches:

- 1. Power down /up
- Use complicated hardware (in BS only)
- 3. Foreign modulation piecewise solution.



Approach 1 is not desirable, but can get away with the transition problem, not really solving the issue.

Approach 2 is expensive, only applicable to BS, and it is far from ideal solution.

Approach 3 might be the best way of the three as long as it works. The fact is that this approach performance vary depends on the GMSK payload and point of process (even or odd symbol index). It introduces unnecessary modulation, which could be a source of interference to others in the network.

Based on our knowledge we are not aware of any one use the proposed approach.

### Brief description of the invention:

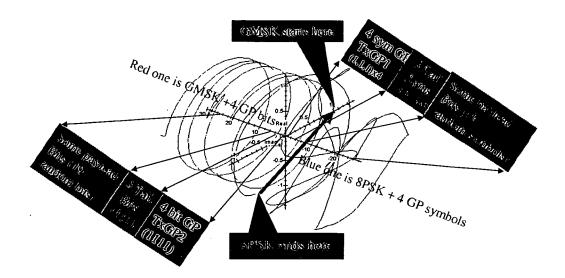
The proposed method is based on in-depth understanding of the details of the two modulations and their behaviour. It considers five factors of the two modulations, the amplitude, the amplitude rate of change, the phase and the rate of phase change and time when these happens. We have searched through all the possible combinations (more than 30 million) of the two modulations, and found groups of the codes at certain moment that all the above factors are in the right condition for smooth transition. The extensive search and post processing enable us to find specific codes and time of switching. The detail is related with the constraint of the HW (the way it produces the two modulations, but the general methodology applies to any two modulators that can produce 8PSK and GMSK).

### <u>Describe how the invention solves the stated problem and achieves the stated</u> advantage(s):

### The problem

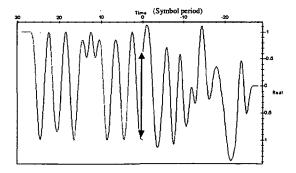
Generally the two modulations bear no relation with each other. They can start and end at any position in the constellation. One typical example is shown below where the two are not in the position to smoothly join each other by switching.





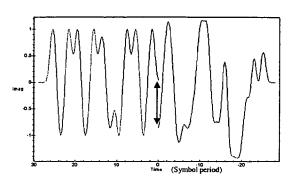
The I and Q observation shown below has obvious steps, which represent high frequency component.

Real: I component

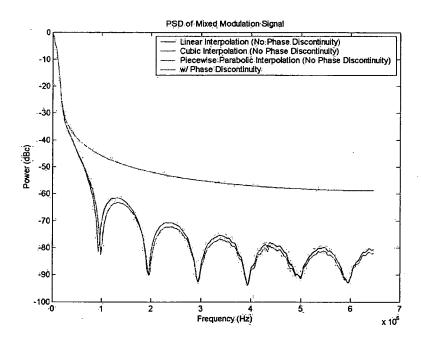




Imagery: Q



The spectrum of this kind of transition is shown below.



The issue is that the two modulations will not meet at the five conditions stated above. The problem of unwanted frequency components will arise if directly make such transition, either through line interpolation or direct switching. What is needed is an adaptive way, suitable for any payload, that bring the end of the two modulation close enough to achieve smooth transition.

### **Proposed resolution**

With the switching codes and control of switching time stated above we can achieve a smooth transition between the two modulations:

Here is a special example that will work for the least flexible HW implementation. It assumes:



Modulation over sample rate (OSR): 48 samples/symbol

Modulation starting phase of GMSK: fixed (0°) Modulation starting phase of 8PSK: fixed (0°)

GMSK switching code: 11111101 8PSK switching code: 06605

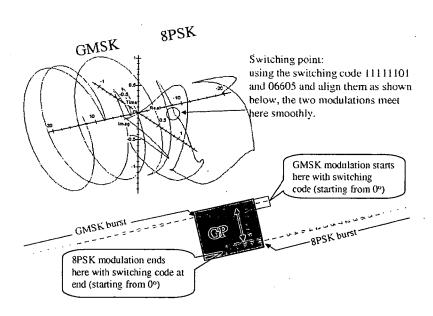
The two modulations signal (not the power) peak level ratio is 8PSK:GMSK = 1.521.

Transition phase: To show how it works as an example, suppose the first burst ends in such a way so that the first symbol of the switching code starts from 0° (For those that starts from none 0 phases due to payload, they can be treated accordingly, simple enough to implement, this is just an example). Make sure the starting time of the next burst modulation is set so that the code in red are aligned with each other. The next burst switch code start from 0 phase.

Switching time: As this example is using OSR48, there will be 48 points for each symbol period. Let's assume that the point right at the beginning of a symbol is indexed with 0 and the point right at the end of the symbol is indexed as 48, or 0 of the next symbol. During the symbol period of the red code, switch right after the 13<sup>th</sup> point.

Result at the point of switching:

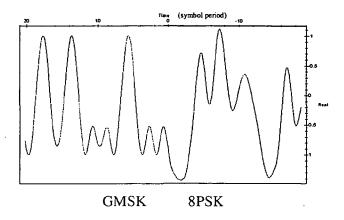
- 1. GMSK amplitude: 1, 8PSK amplitude: 0.999566. Matching well, the difference is less than 1 bit of 12-bit DAC
- 2. Amplitude change rate: GMSK: 0; 8PSK: -0.0166/step: within the range of 8PSK. (C0 allow 0.02/step, OSR48, the modulation would allow more)
- 3. GMSK phase: 24.375°. 8PSK phase:22.583°
- 4. Phase change rate GMSK: 1.875°/step. 8PSK: 0.869°/step.
- 5. Phase changing direction: both are anti-clock wise.



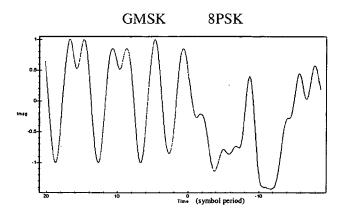
Both I and Q of this approach are smooth as shown below.



Real: I component

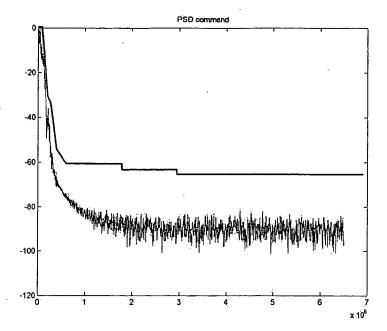


### Imagery: Q component



The spectrum of them is shown below, and as expected, no obvious spurious frequency components were generated from such transition. The read lines are the 8PSK mask defined in 3GPP TS 45.005.





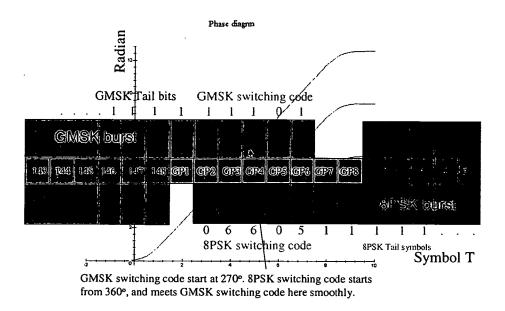
It should be noted that the baseband only provide necessary condition, not the sufficient condition. It must be combined with RF techniques to tackle the issue, including the power level change between the two bursts. The above technique can be used in flexible way, in terms of the codes and switching symbol. The examples here are for demonstration purposes, and in practice it can be different, depends on the RF requirement.

Depend on the code selected; the switching time normally would be different. For general case there are four possible variants for the selected code. The pattern is exactly the same, but with 90° rotation. This is due to the GMSK ending symbol phase is related with its index and the payload. If the modulators start from 0° or 90° or 180° or 270° for the first symbol. The 8PSK->GMSK transition should always be at even symbols counting from the first 8PSK symbol to the switching 8PSK symbol. This is to ensure the switching symbol is on the axis where modulations can start.

Another important factor is the starting phase of the switching code. They should be aligned for the switching code.

As an example here is the illustration of the above example: The read curve shows the GMSK symbol phase from the ending tail bit to the GP, the starting phase can be an arbitrary value, in this case 0°. The switching symbol is arranged to be the fourth GP symbol. We should be able to find the first switching code phase which is 270°. For this set of the code, the 8PSK switching code should be leading by 90°. So we start 8PSK modulation from 360° (or 0°). Then we can switch at the 13<sup>th</sup> OS48 point. They can be easily implemented (sounds more complicated than the actual code).





### What projects/products/standards does it apply to?

Mobile phone, base station and modem of telecomm industry. It can apply to EDGE capable devices.

### How is it going to make us money?

Two ways:

Better way of doing fundamental data transmission in our chip would win more customers. This is one way of making money.

As we know spectrum is expensive and there is strict requirement on the RF resource usage. Everyone in this field sooner or later needs to resolve such issue, and there is no better way to achieve it by changing the codes and applying right time for switching. Others may use this method and pay royalty. This is another way of making money.

### How is it going to give us a competitive edge?

Easier and better way of modulation transition will attract operators to use our solution, as this will have impact on the network performance.

The performance will attract others to use the same kind of transition approach. If we have patent on it and put it in standard, we will be in advantage position on this issue.

### How easy is it to design around?

There are other ways around, but it is going to cost more and introduce foreign modulations. It may not guarantee the performance (generates unwanted spectrum).

### How easy is it to detect when others are using it?

It is very easy to find, as the RF signal over the air will reveal itself what codes and method has been used. With our burst analyser, we can find from baseband raw measurements. Or just measure its code in the guard period and decode them.



### Will it be built into a standard?

Could be but not have to.

As this is not only for MS, but also for BS. The specified method will reduce the interference in the network, hence the performance would be better. Good for generating revenue.

### Give details here of when the idea was conceived, simulated and/or tested: -

Invention conceived on: -	11/09/2003	(MM/ DD/ YYYY)
Simulation completed on: -	26/11/2002	(MM/ DD/ YYYY)
Device Tested on:		(MM/ DD/ YYYY)

### Identify here any related invention disclosures, patent applications and patents:

### **Inventors:**

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### TRANSITION BETWEEN GMSK AND 8PSK BURSTS

A fine tuned interpolation approach....

Zhi-Zhong Yu Philip Children



### Outline

- What is it?
- An adaptive interpolation: the extreme case
- How to find it?
- The procedures of finding the solution
- How does it work?
- A visual example of 8PSK →GMSK
- How about the performance?
- Spectrum analysis and comparison (fixed point)
- How to implement it?
- Simple adaptive implementation and illustrations



## An extreme case of interpolation method...

- As one of the follow up action items of system HLD review, with adaptive approach has been used to achieve smooth transition. method for the transition between GMSK and 8PSK bursts. An Kuei's expert collaboration, we fine tuned the interpolation
  - switching were found to be useful for smooth transition without ntroducing foreign modulations. This is an extreme case of A set of GMSK and 8PSK switching codes, and time of the nterpolation but with 0 interpolation point for OSR48.
- spurious frequency components were found. It proves that it has no more spectrum than those of 8PSK and GMSK. Therefore it With joint effort, the initial spectrum analysis of this transition has been made. As shown by following slides, no obvious could be one of the possible solutions to this issue.



# Four steps to find the switching codes and time

The aim is to find a point in both modulations that

- same amplitude,
- same phase, and
- similar change rate of amplitude and phase.
- More importantly at the same time



# Step 1: Find possible switching point on 8PSK

Search for a 5-symbol-long 8PSK codes that can provide the right amplitude for the transition

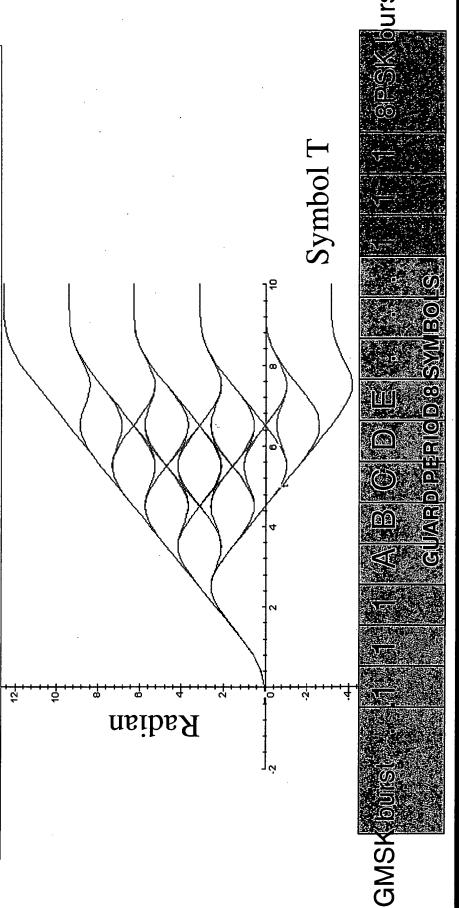
•A list of 8PSK codes were obtained from calculation of 30 million points. Table below shows the code sequence (06605), the OSR index of the middle symbol, amplitude and phase with the difference to the previous OS point.

•Here is one candidate whose switching code is 06605. We are particularly interested in OS point 13 to 14 and its 8PSK phase pattern.

0.977034	0.994412	1.011643	0.933361	0.813868	0.840949	0.869024	0.000100	0.928213	0.959345	0.991504	1.024683	1.058867
-0.009845 dif_arg	-0.00952 dif_arg	-0.009186 dif_arg	-0.012324 dif_arg	-0.017294 dif_arg	-0.017075 dif_arg	-0.016845 dif_arg	नी, निव्यक्ति निव्यक्ति	-0.01635 dif_arg	-0.016083 dif_arg	-0.015803 dif_arg	-0.015509 dif_arg	-0.015199 dif_arg
ji di	dif	dif	ŧ	ŧ	ij	ji Git	<b>₩</b>	ij	ij	ĕ	<del>=</del>	ŧ
17.11954 dif	18.11395 dif	19.1256 dif	20.05896 dif	20.87283 dif	21.71378 dif	22.5828 dif	इस्त्र/रहण्डा वारि	24.40912 dif	25.36847 dif	26.35997 dif	27.38466 dif	28.44352 dif
phase	phase	phase	phase	phase	phase	phase	<b>जिल्ला</b>	phase	phase	eseud	phase	phase
1.098419 phase	1.088899 phase	1.079712 phase	1.067387 phase	1.050093 phase	1.033017 phase	1.016171 phase	මුදුමුදුවේම මූබ්සප	0.983216 phase	0.967132 phase	0.951329 phase	0.93582 phase	0.92062 phase
amp	amp	amp	amp	amp	amp	amp	ভূম্মিট	amp	amp	amp	amp	amp
7	8	6	10	11	12	13	143	15	16	17	18	19
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9 9	9 9	3 6	9 (	9 (	3 6	9 (	6606	9	9 (	3 6	9 9	9 9
0	9	9	9 0	9 0	9	9 0	<b>0</b>	9	9 0	9 0	9	0
[1~5]	[1~5]	ري ح	[1~5]	[1~5]	[1~5]	[1~5]	∭~ব্য ⊹ ©	[1~5]	[1~5]	[1~5]	[1~5]	[1~5]
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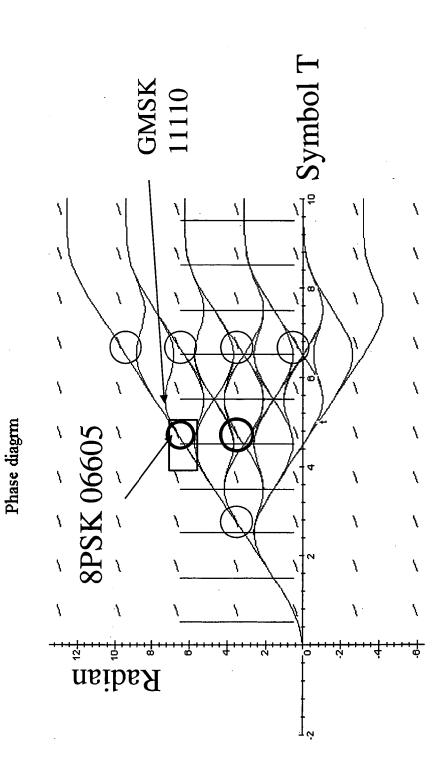
Question: during the period of 13 to 14 as shown above, is there a GMSK modulation point that could smoothly carry on from there? YES....







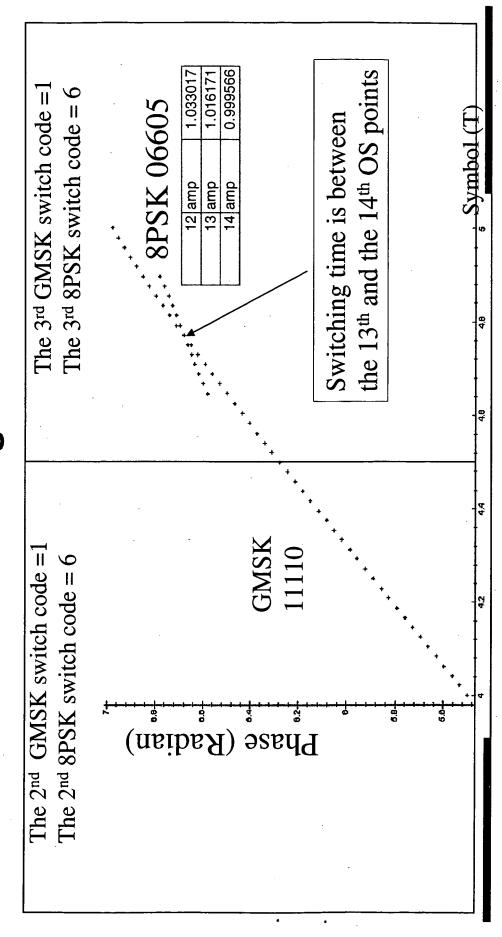
# Step 3: Superimpose "8PSK pattern" on to GMSK traces



7 possible switching points are circled. Let's zoom in



### Step 4: Zoom in of 11110 with 06605 to find switching time to be ..





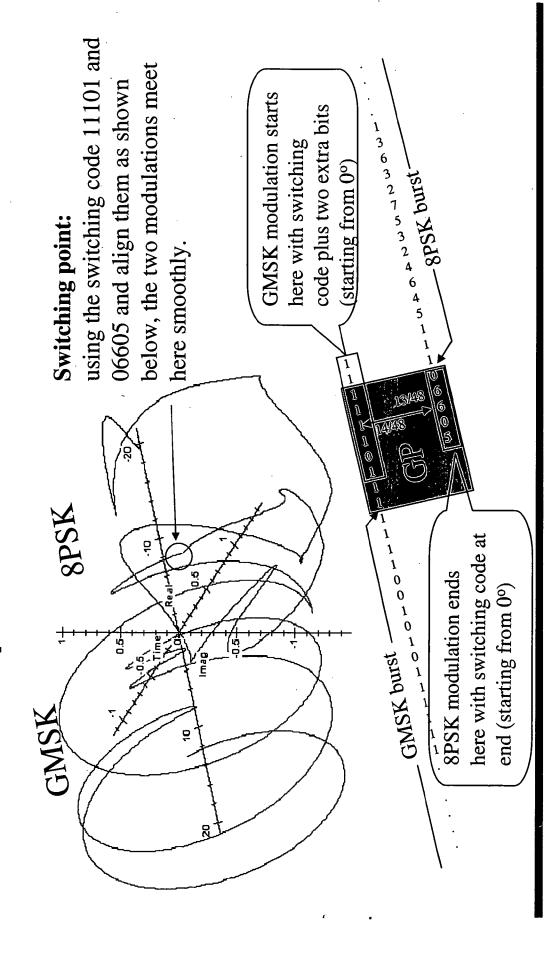
# A visual example of 8PSK →GMSK

Get a feeling of

- How does it works?
- Where is the switching point?
- Does it look smooth?

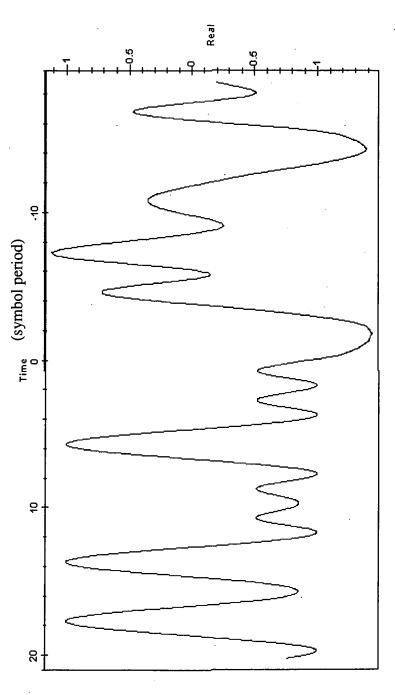
### QUALCOMM (UK) Limited

# An example of 8PSK → GMSK transition





## component: Plot of Real sequence around the 8PSK → GMSK transition

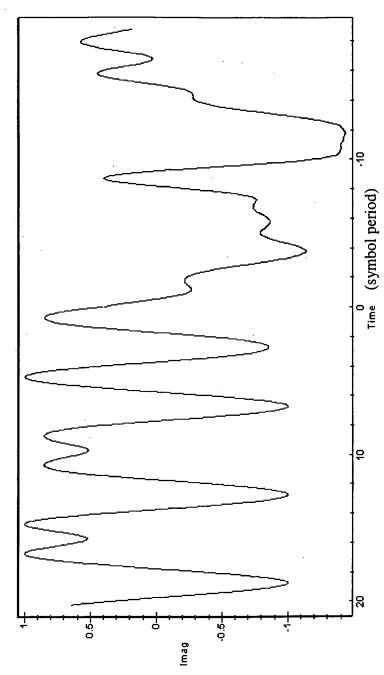


**GMSK** 

8PSK



## **Q** component: Plot of Imagery sequence around 8PSK →GMSK transition GMSK 8PSK





# **Baseband smooth transition**

- them, but also leave enough headroom for RF to play. Therefore transition. This implies that Baseband not only needs to satisfy Baseband smooth transition is a necessary condition not a The spec defines the RF spectrum due to modulation and sufficient condition.
- is to provide smooth signal and provide suitable condition for RF Ramp down/up is a RF activity where the best baseband can do to perform this activity.
- Working on Basedband, it would be important to understand the requirement from RF section, so that baseband provide suitable necessary condition.
- Based on the comments received on switching method, another set of switching codes are demonstrated in the next few slides.



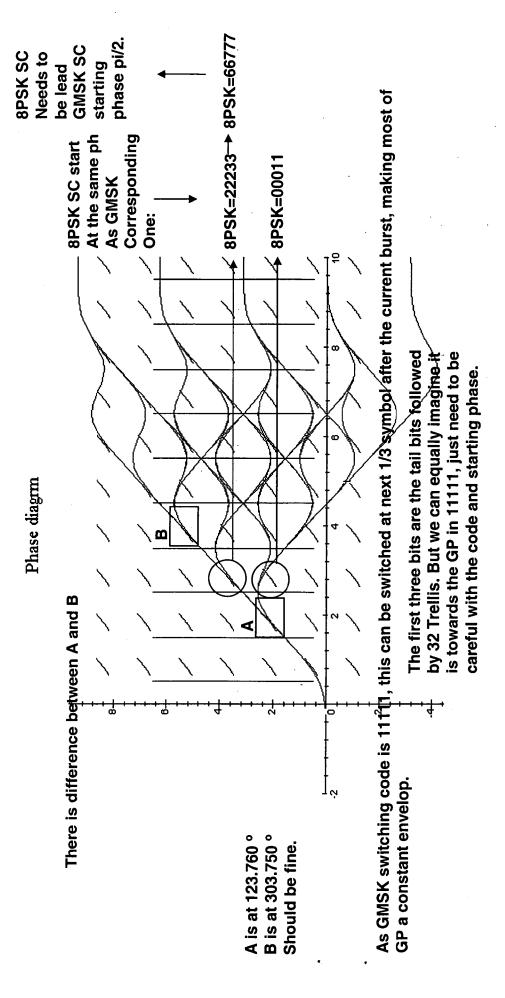
# 8PSK switching code 00011

8 8 8 8	<u> </u>	90 90	ree.	9 9 0	9 9	Jee.
0.00479 dif_arg. 11.289062 degree -0.00423 dif_arg. 11.299709 degree -0.00366 dif_arg. 1.309135 degree -0.00693 dif_arg. 1.289257 degree	269561 dègree. 298051 degree. 326133 degree.	0.01047 dif_arg 1.35367 degree. 0.00988 dif_arg 1.380516 degree. 0.00927 dif arg 1.406519 degree.	0.00865 dif_arg 1.431522 degree. ඔරාවෙග් dif_arg 1.435834 degree.	0.00735 dif_arg 1.477883 degree 0.01414 dif_arg 1.721122 degree	0.01431. dfr.arg 11.868997 degree 0.01327. dfr.arg 11.824068 degree	0.01218 dif_arg: 1'97/081 degree 0.01492 dif_arg≈ 2.035679 degree
8906 9970 0913 8925	6956 9805 2613	3536 8051 0651	3152 5536	7788 2112 1245	2408 2406	77.08 3567
2		 E. 4	1. 4.0	4.7.	 	6 0 0 0
0.00479 dif_arg -0.00423 dif_arg -0.00366 dif_arg -0.00693.dif_arg	-0.01214 dif_arg^1. -0.0116 dif_arg_1. -0.01104 dif_arg_1.	0.01047 dif_arg 0.00988 dif_arg 1 0.00927 dif_arg 1	-0.00865 dif_arg	[arg	0.01431, dif_arg 0.01327, dif_arg	0.01218 dif_arg 0.01492 dif_arg≏
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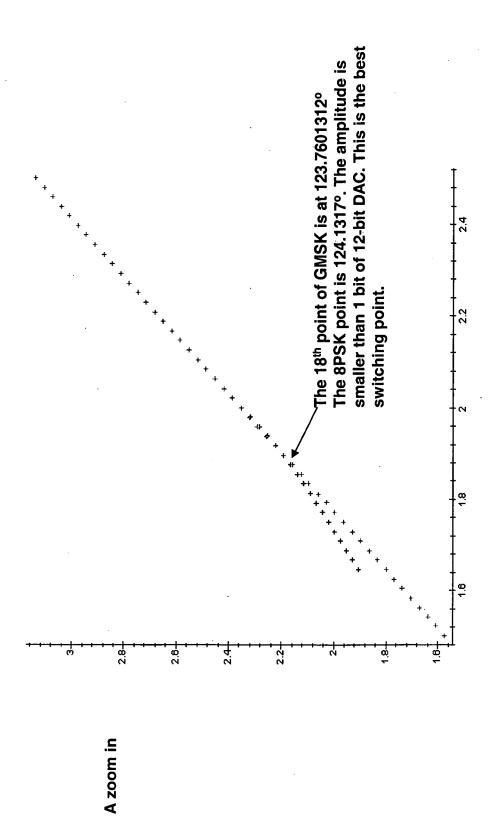
Another set of 8PSK code 00111 or 00011 can be used to make "constant" GP amplitude. It also matches better than 06605. The switch point is at 18th point of the middle symbol 0, or the 30th point of the middle symbol 1.

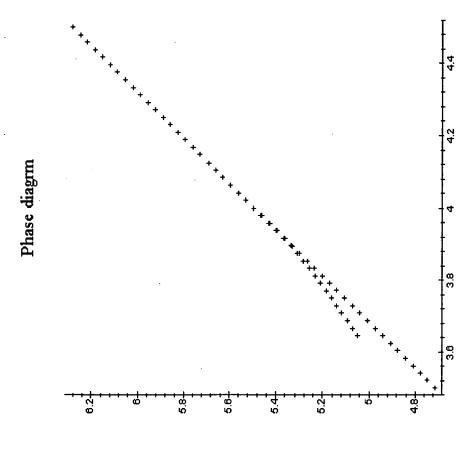


## **GMSK11111 8PSK 00011**









B zoom in

-Symbol T This small ripple is 8PSK 777777... nature. GMSK switching code can be 11111. 8PSK switching code can be 66677, Cannot get 8PSK as constant as GMSK fill GP with 1 to constant envelop 8PSK Tail symbols Amplitude variation in GP GMSK switching code 1 1 1 1 1 8PSK switching code QUALCOMM (UK) Limited JUALCOMM Normalised DAC output

and the switching happens at GP1 to make most GP amplitude constant



## **Question to RF experts**

1. Is this amplitude profile in GP OK?



2. The previous slide shows the amplitude in GP when perform smooth switching from GMSK to 8PSK. This is with an early switching to 8PSK (switching at GP1). For obtaining constant amplitude we can also have a late switch to 8PSK (switching at GP8). Which is preferred? Should we keep GMSK as long as we can?

## The priority might be related with following cases:

and last switching when it is ramp down, so that the switching points are always happen at low power rather than high 2.1 Should the switching point related with ramp down/up? I.e. is it better to have an early switching when it ramp up power?

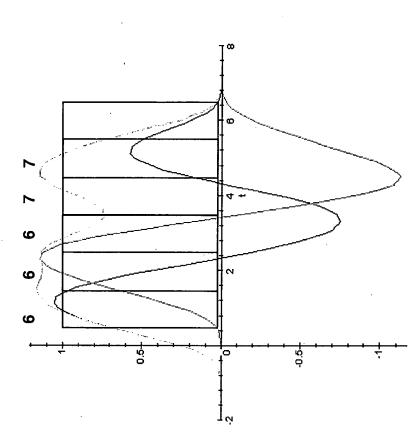
### ŏ

2.2. Should the switching point related to the direction of switching GMSK → 8PSK or opposite direction.

2.3 If the above two are related, then there are four combinations of ramp down/up and directions. In each of the four



# The lower down for 66677



Please note that if it is G28, there would be a step up which should take the switch at 66777, the phase of which need to be carefully studied (have not yet.

For 82G, there is a step down and the switch should be at 66677. This has been studied and know to be at 18 the point of OS.



### Another set of data

- Switch at centre
- Directly from 111 to 777, four cases each has init 90° shift
- Need quick response to the data requirement
- See the switching point:
- » G28: goes high, switching at the 30th point;
- » 82G: goes low, switch at the 18th point.
- Shift from 000111 to 666777, shift by 90° would be suitable for one case of GMSK
- Plan for generating data: look at the GMSK landing



# For matlab analysis 8G8 is better

- If you take extra 2.5 symbols at both ends, then there is a good 0 ends to avoid artificial windowing effect
- For FFT we can also make it G8G and take care of the ending points for FFT to repeat



## Start modulation in 45°

- Condition: Use 11111... and 7777..., and make the switching at GP4,
- There need to be 45° starting phase for GMSK.
- Even we shift switching point to GP3 or GP5, we only get starting phase, as with 45° the next round will come back 22.5° and 67.5° starting phase, which is worse than 45° to axis, while 22.5° will need 67.5°.
- The only possible place that allows axis starting point is GP2 or GP6 switching point.
- These are the options, nothing else under the condition
- For cyclic table, we just need 4 bursts.

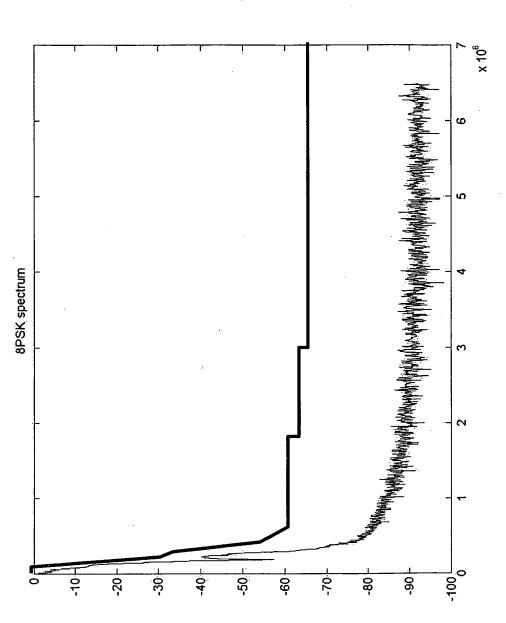


# Spectrum analysis of the transition

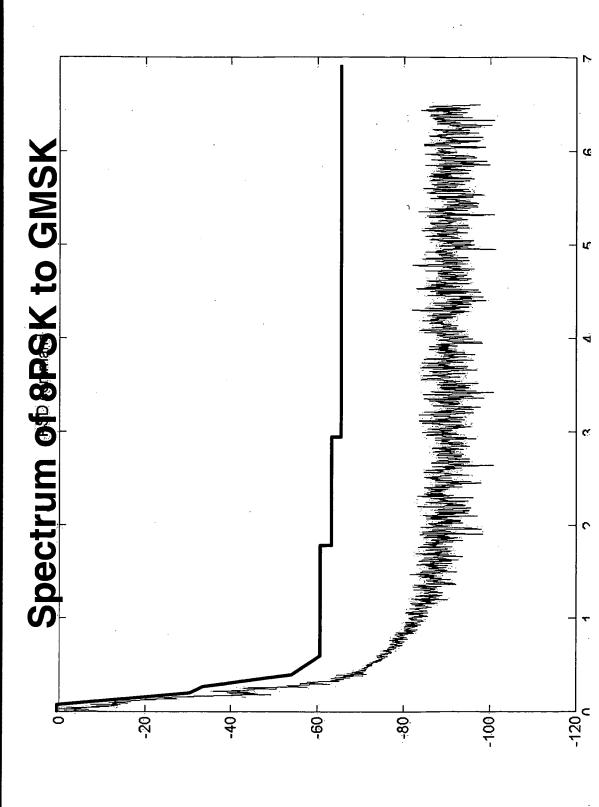
to judge the performance by spectrum analysis. A As the spec defines the spectrum mask, we have analysed and it has been found to be as good as expected - no spurious frequency. In reality it simple and possibly the worst scenario is should be better ...

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# Spectrum for 8PSK only burst









# The spectrum comparison

- spectrums, the spectrum with the smooth transition between 8PSK and Compared with 8PSK only and GMSK only (GMSK not included here) Therefore it is the ideal solution for GMSK and 8PSK transition from spectrum is no more than the combination of the two modulations. frequency components that beyond the spec. It is believed that its GMSK is well below the both masks and has no obvious spurious spectrum point of view.
- random data from 8PSK to GMSK. More bursts with such transition can certain effect at bottom left corner due to window effect of the analysis. be formed to give better and smoother plot than this one. It is believed So far we have made an initial spectrum analysis of two bursts with that the initial spectrum analysis does reflect its true spectrum with
- 8PSK transition spectrum would be the same as 8PSK to GMSK with As GMSK and 8PSK has symmetrical ISI and burst format, GMSK to this method
- As it satisfies the spectrum during useful part of the burst, there is no doubt that it will satisfy the transition spectrum of either GMSK or



### **Implementation**

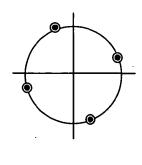
- Decide the symbol of switching in GP
- switching symbol with the correct switching code starting Use the switching code two symbols earlier than the phase
- Switching between the 13th and 14th OSR48 points



# Four possible switching points

there are four possible switching points due to the GMSK modulation payload and the symbol index In general case, with GMSK switch code 11101, formed by 90° rotation of GMSK modulation. of the switching symbol. The four points are

rotations, keeping 06605 pattern. GMSK switching switching code can be obtained with the same code starting phase should lag 8PSK by 90°. To meet each of the four points, the 8PSK





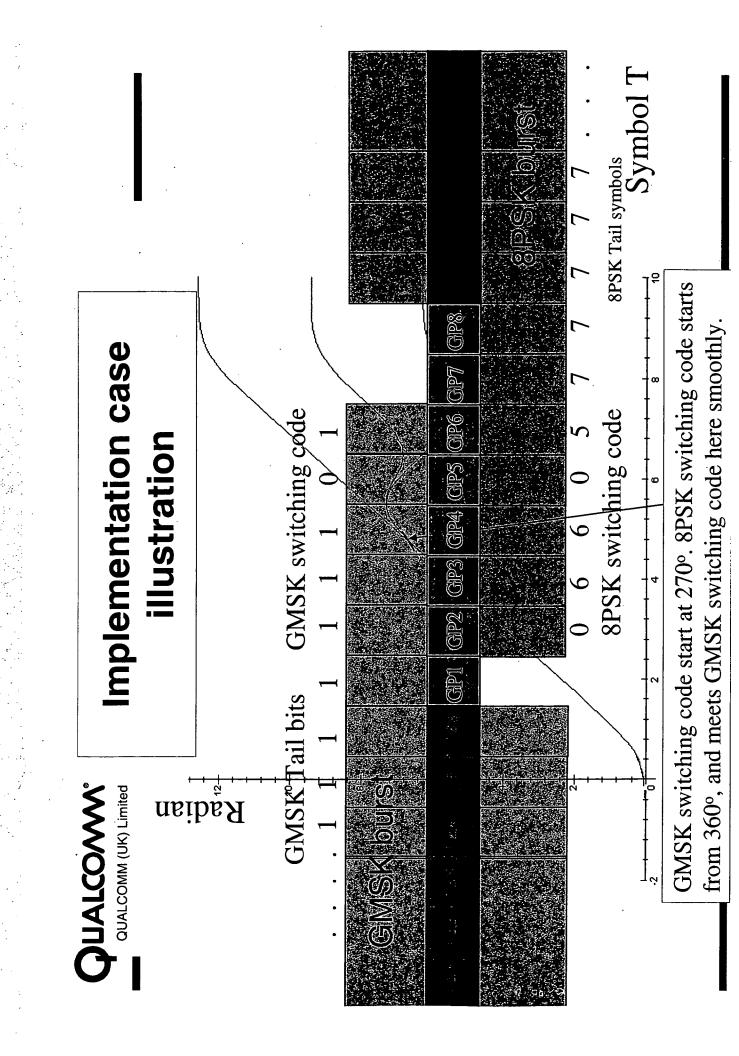
### slightly bigger than the baseband phase at For GMSK 11111 the symbol phase is last moment of the symbol period

This could be used to tell where the symbol is and work out the starting phase of the other it switches to.



# Implementation case -- an example

- oossible approach based on individually implemented GMSK and 8PSK modulators: (all angles mentioned below refer to symbol, not baseband Practical implementation depends on modulators design. Below is one
- Due to the need for switching, both modulations run simultaneously in GP.
- Suppose that modulation starts from 0° or 90° or 180° or 270° for the first symbol, and we are using OSR48. We switch between the 13<sup>th</sup> and 14<sup>th</sup> point of the 4<sup>th</sup> GP symbol, similar to what has been demonstrated
- 11101, filling 1 bit after the tail bit forms 1111101; the 8PSK switching code 06605 pattern can start from the 2nd GP symbol with 90° leading phase of For GMSK to 8PSK, suppose that the chosen GMSK switching code is GMSK switching code.
- For 8PSK to GMSK, we need to make 8PSK switching code 06605 pattern from  $90^{\circ}$  . At the same time start GMSK modulation with 11101 from  $0^{\circ}$ . start on even symbol of current modulation, with the first code that start





### Conclusions

- A novel baseband modulation transition method has been demonstrated.
- It has smooth transition between consecutive 8PSK and GMSK bursts (in any order) without introducing interpolation points other than themselves for OSR48.
- The transition spectrum is no more than those of 8PSK and GMSK. Therefore it is the optimum transition method from spectrum point of view.
- JL chain, which is important to satisfy the ETSI spec with good This enables the baseband to take near 0 budget for the entire performance.
- As demonstrated, it could be one of the best solutions for the modulation transition issues.
- It is low risk and no compromise approach.